

APPLICATION NOTE

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Voltage-Controlled Amp Covers 55 dB Range

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INTRODUCTION

By using a dual multiplier chip and a dual high-speed op amp (Figure 1), you can build a 2-chip voltage-controlled amplifier with a dynamic range of 55 dB, a 3-dB bandwidth of 8 MHz, and exponential control. The amplifier's output ranges from 5 mV p-p at $V_X = 0 \text{ V}$ to 3 V p-p at $V_X = 3 \text{ V}$ for a 100 Ω load. The circuit's gain is unity at V_X = 2 V. You can also use Figure 1 to drive a reverse-terminated 50 Ω cable to 1.5 V p-p. Or you can use each multiplier-op amp combination separately to amplify two signals with control by a common voltage.

Figure 1's circuit connects the AD539's two voltage-in current-out multipliers in series. Each of the op amps acts as a current-to-voltage converter. V_X , a single 0 V to 3 V dc input, controls both multipliers. Because both multipliers are in series, the overall transfer function is

$$\frac{V_{OUT}}{V_{IN}} = \frac{V_X^2}{4 V^2}$$

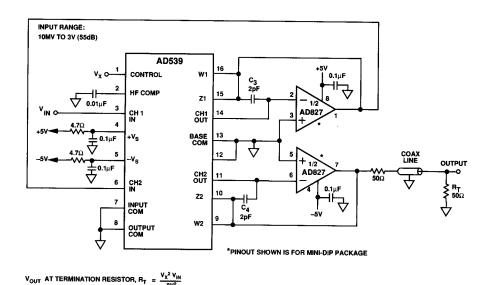


Figure 1. A Wide Range Voltage-Controlled Amplifier Circuit

 V_{OUT} AT PIN & OF AD827 = $\frac{V_{\chi^2} V_{IN}}{4V^2}$

The plot of V_x vs. the gain of this voltage-controlled amplifier on log-log axes is a straight line, which demonstrates the exponential gain response.

The square term in the denominator of the transfer function comes from connecting each of the multiplier's W and Z outputs. The W and Z pins of the AD539 are each connected to 6 k Ω resistors. Connecting the two pins sets the two resistors in parallel and thus halves the gain. The feedback resistor in each current-to-voltage converter halves from 6 k Ω to 3 k Ω and thereby reduces the amplifer's overall gain by a factor of four.

As an option to Figure 1's circuit, you can disconnect the Z outputs and use only the W outputs, thereby fixing the gain resistor at 6 k Ω . If so, the overall transfer function is

$$\frac{V_{OUT}}{V_{IN}} = \frac{V_X^2}{1 V^2}$$

The maximum gain is 9 when $V_X = 3 V$. You can trade decreased bandwidth for increased gain by adding an external scaling resistor, Rs, in series with the on-chip feedback resistors. In this case, the transfer function for the dual-multiplier circuit becomes

$$\frac{V_{OUT}}{V_{IN}} = \frac{V_X^2}{1 \ V^2} \left(\frac{R_S}{5R_S + 6.25} \right)^2$$

where the units of R_s is $k\Omega s$.